# Low-Power Video Switches for Dual SCART Connectors 

## General Description

The MAX9655/MAX9656 dual SCART switches route video signals between a set-top box decoder chip and two external SCART connectors. Under the control of the TV_SEL logic input, the MAX9655 selects whether the CVBS and RGB signals from the encoder or the VCR SCART are routed to the TV SCART. The CVBS signal from the encoder is always routed to the VCR SCART.

The MAX9656 is similar to the MAX9655 except that under the control of the VCR_SEL logic input, the MAX9656 selects whether the CVBS signal from the encoder or the TV SCART is routed to the VCR SCART. The MAX9656 also features a low-power shutdown mode, in which quiescent current falls to $35 \mu \mathrm{~A}$.
The incoming video signals must be AC-coupled to the inputs, which have sync-tip clamps to set the internal DC level. After the input stages, multiplexers select which video signals are routed to the reconstruction filters and output amplifiers. The reconstruction filters are optimized for standard-definition signals and typically have $\pm 1 \mathrm{~dB}$ passband flatness out to 9.5 MHz and 47 dB attenuation at 27 MHz .
The amplifiers have 2V/V gain, and the outputs can be DC-coupled to a $75 \Omega$ load, which is the equivalent of two video loads, or AC-coupled to a $150 \Omega$ load.

## Applications

SCART Set-Top Boxes

| - Dual SCART Support for Video Signals |  |  |
| :---: | :---: | :---: |
| - Supports CVBS Input from TV SCART (MAX9656) |  |  |
| - Reconstruction Filters with 9.5MHz Passband and 47 dB Attenuation at 27 MHz |  |  |
| - Fixed Gain of 2V/V |  |  |
| - Input Sync-Tip Clamps |  |  |
| - 2.7V to 3.6V Single-Supply Operation |  |  |
| Ordering Information |  |  |
| PART | PIN-PACKAGE | TV SCART CVBS RETURN SUPPORT |
| MAX9655AEE+ | 16 QSOP | No |
| MAX9656AEP+ | 20 QSOP | Yes |

Note: All devices are specified over the $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$ operating temperature range.
+Denotes a lead-free/RoHS-compliant package.

Pin Configurations and Typical Application Circuits appear at end of data sheet.

Functional Diagram


## Low-Power Video Switches <br> for Dual SCART Connectors

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage

| VDD to GND .-0.3V to +4 |  |
| :---: | :---: |
| All Video | 0.3 V ) to +4V |
| Duration of Output Short Circuit to VDD or GND |  |
| Continuous Input Current |  |
| All Video and |  |
| Continuous Power Dissipation ( $\mathrm{T}_{\mathrm{A}}=+70^{\circ} \mathrm{C}$ ) |  |
| 16-Pin QSOP (derate $8.3 \mathrm{~mW} /{ }^{\circ} \mathrm{C}$ above $+70^{\circ} \mathrm{C}$ ) ........... 667 mW |  |
| -Pin QSOP |  |

Operating Temperature Range .......................... $-40^{\circ} \mathrm{C}$ to $+125^{\circ} \mathrm{C}$
Junction Temperature ..................................................... $+150^{\circ} \mathrm{O}$
Storage Temperature Range ............................. $65^{\circ} \mathrm{C}$ to $+150^{\circ} \mathrm{C}$
Lead Temperature (soldering, 10s) ................................. $+300^{\circ} \mathrm{O}$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability

## ELECTRICAL CHARACTERISTICS

$\left(V_{D D}=3.3 V, V_{G N D}=0, V_{\overline{S H D N}}=V_{D D}, V C R \_S E L=V_{D D}, T V \_S E L=V_{D D}, R_{L}=150 \Omega\right.$ to $G N D, T_{A}=T_{M I N}$ to $T_{M A X}$, unless otherwise noted. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.) (Note 1)

| PARAMETER | SYMBOL | CONDITIONS |  | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply Voltage Range | VDD | Guaranteed by power-supply rejection test |  | 2.7 | 3.3 | 3.6 | V |
| Quiescent Supply Current | IDD | No load |  |  | 21 | 45 | mA |
| Shutdown Supply Current | ISHDN | $\begin{aligned} & \text { VSHDN }=\text { TV_SEL }=\text { VCR_SEL }=\text { GND } \\ & \text { (MAX9656 only) } \end{aligned}$ |  |  | 35 | 70 | $\mu \mathrm{A}$ |
| Input Voltage |  | Unselected input |  |  | VDD/3 |  | V |
| Input Resistance |  | Unselected input |  | 222 |  |  | $\mathrm{k} \Omega$ |
| Sync-Tip Clamp Level | VCLP |  |  | 0.23 | 0.3 | 0.39 | V |
| Input Voltage Range |  | Guaranteed by outputvoltage swing | $2.7 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq 3.6 \mathrm{~V}$ |  |  | 1.05 | VP-P |
|  |  |  | $3.0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq 3.6 \mathrm{~V}$ |  |  | 1.2 |  |
| Sync Crush |  | Sync-tip clamp; percentage reduction in sync pulse ( 0.3 V P-P); guaranteed by input clamping current measurement |  |  |  | 2 | \% |
| Input Clamping Current |  |  |  |  | 1 | 2 | $\mu \mathrm{A}$ |
| Maximum Input Source Resistance |  |  |  |  | 300 |  | $\Omega$ |
| DC Voltage Gain | Av | $\begin{aligned} & \text { RL=150 } \text { to GND } \\ & \text { (Note 2) } \end{aligned}$ | $\begin{aligned} & \mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}, 0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{IN}} \\ & \leq 1.05 \mathrm{~V} \end{aligned}$ | 1.96 | 2 | 2.04 | V/V |
|  |  |  | $\begin{aligned} & V_{D D}=3.0 \mathrm{~V}, 0 \mathrm{~V} \leq \mathrm{V}_{\mathrm{IN}} \\ & \leq 1.2 \mathrm{~V} \end{aligned}$ | 1.96 | 2 | 2.04 |  |
| DC Gain Mismatch |  | Guaranteed by output-voltage swing |  | -2 |  | +2 | \% |
| Output Level |  | Measured at output, $\mathrm{Cl}_{\text {IN }}=0.1 \mu \mathrm{~F}$ to GND |  | 0.218 | 0.3 | 0.39 | V |
| Output-Voltage Swing |  | Measured at output, $\mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}, \mathrm{~V}_{\mathrm{IN}}=$ $V_{C L P}$ to ( $V_{C L P}+1.05 \mathrm{~V}$ ), $\mathrm{R}_{\mathrm{L}}=150 \Omega$ to -0.2 V |  |  | 2.1 |  | VP-P |
|  |  | Measured at output, $\mathrm{V}_{\mathrm{DD}}=2.7 \mathrm{~V}, \mathrm{~V}_{\mathrm{IN}}=$ $\mathrm{V}_{\mathrm{CLP}}$ to ( $\mathrm{V}_{\mathrm{CLP}}+1.05 \mathrm{~V}$ ), $\mathrm{R}_{\mathrm{L}}=150 \Omega$ to $\mathrm{V}_{\mathrm{DD}} / 2$ |  |  | 2.1 |  |  |
|  |  | Measured at output, $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}, \mathrm{~V}_{\text {IN }}=$ $V_{C L P}$ to ( $V_{C L P}+1.2 \mathrm{~V}$ ), $\mathrm{R}_{\mathrm{L}}=150 \Omega$ to -0.2 V |  |  | 2.4 |  |  |
|  |  | Measured at output, $\mathrm{V}_{\mathrm{DD}}=3.0 \mathrm{~V}, \mathrm{~V}$ IN $=$ $V_{C L P}$ to ( $V_{C L P}+1.2 \mathrm{~V}$ ), $R_{L}=150 \Omega$ to $V_{D D} / 2$ |  |  | 2.4 |  |  |
|  |  | Measured at output, $\mathrm{V}_{\mathrm{DD}}=3.135 \mathrm{~V}, \mathrm{~V}_{\text {IN }}=$ $V_{C L P}$ to ( $\mathrm{V}_{\mathrm{CLP}}+1.05 \mathrm{~V}$ ), $\mathrm{R}_{\mathrm{L}}=75 \Omega$ to -0.2 V |  |  | 2.1 |  |  |

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## ELECTRICAL CHARACTERISTICS (continued)

 noted. Typical values are at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$.) (Note 1)

| PARAMETER | SYMBOL | CONDITIONS |  | MIN | TYP | MAX | UNITS |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Output Short-Circuit Current |  | Short to GND (sourcing) |  | 140 |  |  | mA |
|  |  | Short to V ${ }_{\text {DD }}$ (sinking) |  | 70 |  |  |  |
| Output Resistance | Rout | $\mathrm{V}_{\text {OUT }}=1.5 \mathrm{~V},-10 \mathrm{~mA} \leq \mathrm{I}_{\text {LOAD }} \leq+10 \mathrm{~mA}$ |  | 0.2 |  |  | $\Omega$ |
| Power-Supply Rejection Ratio |  | $2.7 \mathrm{~V} \leq \mathrm{V}_{\mathrm{DD}} \leq 3.6 \mathrm{~V}$ |  | 48 | 64 |  | dB |
|  |  | $\mathrm{f}=1 \mathrm{MHz}, 100 \mathrm{mV}$ P-P |  |  | 20 |  |  |
| Standard-Definition <br> Reconstruction Filter |  | $V_{\text {OUT }}=2 V_{P-P}$, reference frequency is $100 \mathrm{kHz}, \pm 1 \mathrm{~dB}$ passband flatness |  | 9.5 |  |  | MHz |
|  |  | $\mathrm{V}_{\text {OUT }}=2 \mathrm{~V}_{\text {P-P }}$, reference frequency is 100 kHz | $\mathrm{f}=5.5 \mathrm{MHz}$ |  | 0.1 |  | dB |
|  |  |  | $f=9.5 \mathrm{MHz}$ |  | -1 |  |  |
|  |  |  | $\mathrm{f}=10 \mathrm{MHz}$ |  | -3 |  |  |
|  |  |  | $\mathrm{f}=27 \mathrm{MHz}$ |  | -47 |  |  |
| Differential Gain | DG | 5-step modulated staircase of 129 mV step size and 286 mV peak-to-peak subcarrier amplitude, $\mathrm{f}=4.43 \mathrm{MHz}$ |  | 0.4 |  |  | \% |
| Differential Phase | DP | 5-step modulated staircase of 129 mV step size and 286 mV peak-to-peak subcarrier amplitude, $f=4.43 \mathrm{MHz}$ |  | 0.45 |  |  | deg |
| Group-Delay Distortion |  | $100 \mathrm{kHz} \leq \mathrm{f} \leq 5 \mathrm{MHz}$, outputs are 2VP-P |  | 9 |  |  | ns |
| Peak Signal to RMS Noise |  | $100 \mathrm{kHz} \leq \mathrm{f} \leq 5 \mathrm{MHz}$ |  | 71 |  |  | dB |
| 2T Pulse Response |  | $2 \mathrm{~T}=200 \mathrm{~ns}$ |  | 0.2 |  |  | K\% |
| 2T Bar Response |  | $2 \mathrm{~T}=200 \mathrm{~ns}$; bar time is $18 \mu \mathrm{~s}$; the beginning $2.5 \%$ and the ending $2.5 \%$ of the bar time are ignored |  | 0.2 |  |  | K\% |
| 2T Pulse-to-Bar K Rating |  | $2 \mathrm{~T}=200 \mathrm{~ns}$; bar time is $18 \mu \mathrm{~s}$; the beginning $2.5 \%$ and the ending $2.5 \%$ of the bar time are ignored |  | 0.3 |  |  | K\% |
| Nonlinearity |  | 5-step staircase |  | 0.1 |  |  | \% |
| Output Impedance |  | $\mathrm{f}=5.5 \mathrm{MHz}$ |  | 8.07 |  |  | $\Omega$ |
| All-Hostile Crosstalk |  | $\mathrm{f}=15 \mathrm{kHz}$ |  | -82 |  |  | dB |
|  |  | $\mathrm{f}=4.43 \mathrm{MHz}$ |  | -78 |  |  |  |
| Output-to-Input Crosstalk |  | $\mathrm{f}=30 \mathrm{MHz}$ |  | -68 |  |  | dB |
| LOGIC SIGNALS (MAX9655: TV_SEL, MAX9656: TV_SEL, VCR_SEL, $\overline{\text { SHDN }}$ ) |  |  |  |  |  |  |  |
| Logic-Low Threshold | VIL | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  |  | $\begin{aligned} & 0.3 x \\ & V_{D D} \end{aligned}$ |  | V |
| Logic-High Threshold | $\mathrm{V}_{\mathrm{IH}}$ | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  | $\begin{aligned} & 0.7 x \\ & V_{D D} \end{aligned}$ |  |  | V |
| Logic Input Current | IIN | $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$ |  |  |  | 10 | $\mu \mathrm{A}$ |

Note 1: All devices are $100 \%$ production tested at $\mathrm{T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}$. Specifications over temperature limits are guaranteed by design.
Note 2: Voltage gain (Av) is a two-point measurement in which the output-voltage swing is divided by the input-voltage swing.

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$$
\left(\mathrm{V}_{\mathrm{DD}}=3.3 \mathrm{~V}, \mathrm{~V}_{\mathrm{GND}}=0, \mathrm{~V}_{\mathrm{SHDN}}=\mathrm{V}_{\mathrm{DD}}, \mathrm{R}_{\mathrm{L}}=150 \Omega \text { to } \mathrm{GND}, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C} .\right)
$$



LARGE-SIGNAL GAIN
vs. FREQUENCY


VIDEO CROSSTALK
vs. FREQUENCY


Typical Operating Characteristics


LARGE-SIGNAL GAIN FLATNESS vs. FREQUENCY


VIDEO GROUP DELAY DISTORTION vs. FREQUENCY


## Low-Power Video Switches for Dual SCART Connectors

## Typical Operating Characteristics (continued)

$\left(V_{D D}=3.3 \mathrm{~V}, \mathrm{~V}_{G N D}=0, \mathrm{~V}_{\overline{S H D N}}=\mathrm{V}_{\mathrm{DD}}, R_{\mathrm{L}}=150 \Omega\right.$ to $\left.G N D, \mathrm{~T}_{\mathrm{A}}=+25^{\circ} \mathrm{C}.\right)$


DIFFERENTIAL GAIN






DIFFERENTIAL GAIN


## Low-Power Video Switches for Dual SCART Connectors




SYNC-TIP CLAMP CURRENT
vs. INPUT VOLTAGE



QUIESCENT SUPPLY CURRENT
vs. TEMPERATURE


# Low-Power Video Switches for Dual SCART Connectors 

Pin Description

| PIN |  | NAME | FUNCTION |
| :---: | :---: | :---: | :---: |
| MAX9655 | MAX9656 |  |  |
| 1 | 2 | ENC_B_IN | Encoder Blue Video Input. AC-couple the signal through a $0.1 \mu \mathrm{~F}$ capacitor. |
| 2 | 3 | ENC_G_IN | Encoder Green Video Input. AC-couple the signal through a $0.1 \mu \mathrm{~F}$ capacitor. |
| 3 | 4 | ENC_R_IN | Encoder Red Video Input. AC-couple the signal through a $0.1 \mu \mathrm{~F}$ capacitor. |
| 4 | 5 | ENC_CVBS_IN | Encoder Composite Video Input. AC-couple the signal through a $0.1 \mu \mathrm{~F}$ capacitor. |
| 5 | 6 | VCR_B_IN | VCR SCART Blue Video Input. AC-couple the signal through a $0.1 \mu \mathrm{~F}$ capacitor. |
| 6 | 7 | VCR_G_IN | VCR SCART Green Video Input. AC-couple the signal through a $0.1 \mu \mathrm{~F}$ capacitor. |
| 7 | 8 | VCR_R_IN | VCR SCART Red Video Input. AC-couple the signal through a $0.1 \mu \mathrm{~F}$ capacitor. |
| 8 | 9 | VCR_CVBS_IN | VCR SCART Composite Video Input. AC-couple the signal through a $0.1 \mu \mathrm{~F}$ capacitor. |
| 9 | 11 | GND | Ground |
| 10 | 13 | VCR_CVBS_OUT | VCR SCART Composite Video Output. The sync tip is biased at 0.3V. |
| 11 | 14 | TV_CVBS_OUT | TV SCART Composite Video Output. The sync tip is biased at 0.3V. |
| 12 | 15 | TV_R_OUT | TV SCART Red Video Output. The sync tip is biased at 0.3V. |
| 13 | 16 | TV_G_OUT | TV SCART Green Video Output. The sync tip is biased at 0.3V. |
| 14 | 17 | TV_B_OUT | TV SCART Blue Video Output. The sync tip is biased at 0.3V. |
| 15 | 18 | TV_SEL | TV SCART Output Selection. Connect to GND to route the encoder video signals to the TV SCART outputs. Connect to VDD to route the VCR SCART video signals to the TV SCART outputs. |
| 16 | 19 | VDD | Positive Power Supply. Bypass with $0.1 \mu \mathrm{~F}$ ceramic capacitors to GND. |
| - | 1 | TV_CVBS_IN | Television SCART Composite Video Input. AC-couple the signal through a $0.1 \mu \mathrm{~F}$ capacitor. |
| - | 10 | N.C. | No Connection. Not internally connected. |
| - | 12 | $\overline{\text { SHDN }}$ | Active-Low Shutdown Logic Input. Connect to GND to place device in shutdown. Connect to VDD for normal operation. |
| - | 20 | VCR_SEL | VCR SCART Output Selection. Connect to GND to route ENC_CVBS_IN to the VCR SCART CVBS output. Connect to VDD to route TV_CVBS_IN to the VCR SCART CVBS output. |

## Detailed Description

A MAX9655 or a MAX9656 can comprise the video portion of a low-cost, dual SCART solution in set-top boxes with a subset of the full SCART functions. The MAX9655/MAX9656 select whether the CVBS, red, green, and blue video signals from the encoder or the VCR SCART are routed to the TV SCART. The MAX9655/MAX9656 support the output of one CVBS signal to the VCR SCART. In the MAX9655, the CVBS signal from the encoder is routed to the VCR SCART.
In the MAX9656, the CVBS signal routed to the VCR SCART can come from the encoder or TV SCART. In the typical usage case, the VCR (or DVD recorder) records a television program from the set-top box. In such a case, the encoder would be the source of the

CVBS signal. Support for the TV SCART CVBS return path is useful when a person wants to record on his VCR (or more likely DVD recorder) a television program received through the television's antenna. The television program is transmitted from the television to the set-top box and then to the VCR.
Both the MAX9655 and MAX9656 have integrated reconstruction filters so that when the encoder video signals are routed to the TV SCART or the VCR SCART, the steps and spikes left by the video digital-to-analog converter (DAC) are smoothed away. Although the incoming video signals from the VCR SCART are assumed to be filtered already, the reconstruction filter has wide enough bandwidth so that the video signals from the VCR SCART are not degraded.

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The incoming video signals can have any DC bias because the input sync-tip clamps restore the DC level. The output amplifiers have a gain of $2 \mathrm{~V} / \mathrm{V}$. The MAX9655/MAX9656 operate from a single 3.3V supply and consume low quiescent power and low average power. In addition, the MAX9656 also has shutdown mode.

## Operating Modes

TV_SEL controls whether the encoder or VCR video signals are sent to the TV SCART. See Table 1.
On the MAX9656, VCR_SEL controls whether the CVBS signal from the TV SCART or the encoder is sent to the VCR SCART. SHDN controls whether the device is on or off. See Tables 2 and 3. In shutdown, the outputs of the MAX9656 are high impedance.

## Input

Every video signal must be AC-coupled to the MAX9655/MAX9656 through $0.1 \mu \mathrm{~F}$ capacitors. The MAX9655/MAX9656 have sync-tip clamps and bias circuits to restore the DC level of the video signal after the input coupling capacitor. When a video input is selected, the input has a sync-tip clamp, which accepts video signals that have sync pulses or that reach their minimum level during sync. Composite video with blanking and sync (CVBS) is an example of a video signal with sync pulses. The red, green, and blue signals in an RGBS signal set are examples of signals that return to their blank level during sync. The sync-tip voltage is internally set to 300 mV .
When a video input is not selected, the inputs to the MAX9655 and the MAX9656 do not distort the video signal in case the video source is driving video signals to another video circuit such as a video multiplexer. The inputs are biased at $V_{D D} / 3$, which is sufficiently above ground so that the ESD diodes never forward bias as the video signal changes. The input resistance

Table 1. TV_SEL Logic (Applicable to Both the MAX9655 and the MAX9656)

| LOGIC STATE | MODE |
| :---: | :--- |
| Low | Encoder video signals are routed to the <br> TV SCART. |
| High | VCR SCART video signals are routed to <br> the TV SCART. |

## Table 2. VCR_SEL Logic (Only Applicable to the MAX9656)

| LOGIC STATE | MODE |
| :---: | :--- |
| Low | CVBS signal from encoder is routed to the <br> VCR SCART. |
| High | CVBS signal from the TV SCART is routed <br> to the VCR SCART. |

Table 3. $\overline{\text { SHDN }}$ Logic (Only Applicable to the MAX9656)

| LOGIC STATE | MODE |
| :---: | :---: |
| Low | Off |
| High | On |

is $220 \mathrm{k} \Omega$, which presents negligible loading on the video current DAC. The sole exception to this condition is ENC_CVBS_IN (MAX9655), in which the input circuit is always a sync-tip clamp. Table 4 summarizes which input circuit is active on the inputs of the MAX9655 depending on TV_SEL. Table 5 summarizes which input circuit is active on the inputs of the MAX9656 depending on TV_SEL and VCR_SEL.

## Table 4. MAX9655 Input Circuit of Input as Determined by State of TV_SEL

| INPUT | INPUT CIRCUIT <br> (TV_SEL = LOW) | INPUT CIRCUIT <br> (TV_SEL = HIGH) |
| :---: | :---: | :---: |
| ENC_B_IN | Sync-tip clamp | Bias |
| ENC_G_IN | Sync-tip clamp | Bias |
| ENC_R_IN | Sync-tip clamp | Bias |
| ENC_CVBS_IN | Sync-tip clamp | Sync-tip clamp |
| VCR_B_IN | Bias | Sync-tip clamp |
| VCR_G_IN | Bias | Sync-tip clamp |
| VCR_R_IN | Bias | Sync-tip clamp |
| VCR_CVBS_IN | Bias | Sync-tip clamp |

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Table 5. MAX9656 Input Circuit of Input as Determined by State of TV_SEL

| INPUT | INPUT CIRCUIT (TV_SEL = LOW) |  | INPUT CIRCUIT (TV_SEL = HIGH) |  |
| :---: | :---: | :---: | :---: | :---: |
| ENC_B_IN | Sync-tip clamp |  | Bias |  |
| ENC_G_IN | Sync-tip clamp |  | Bias |  |
| ENC_R_IN | Sync-tip clamp |  | Bias |  |
| ENC_CVBS_IN | Sync-tip clamp (VCR_SEL = 0) | Sync-tip clamp (VCR_SEL = 1) | Sync-tip clamp (VCR_SEL = 0) | $\begin{gathered} \text { Bias } \\ (\text { VCR_SEL }=1) \end{gathered}$ |
| TV_CVBS_IN | $\begin{gathered} \text { Bias } \\ (\text { VCR_SEL }=0) \end{gathered}$ | Sync-tip clamp (VCR_SEL = 1) | $\begin{gathered} \text { Bias } \\ (\text { VCR_SEL }=0) \end{gathered}$ | Sync-tip clamp (VCR_SEL = 1) |
| VCR_B_IN | Bias |  | Sync-tip clamp |  |
| VCR_G_IN | Bias |  | Sync-tip clamp |  |
| VCR_R_IN | Bias |  | Sync-tip clamp |  |
| VCR_CVBS_IN | Bias |  | Sync-tip clamp |  |

Note: VCR_SEL = X (don't care), except where noted.

When the MAX9656 is in shutdown, its inputs are biased at the same voltage and present the same input resistance as unselected inputs.

## Video Filter

The MAX9655/MAX9656 video filter features $\pm 1 \mathrm{~dB}$ passband out to 9.5 MHz and 47 dB attenuation at 27 MHz , making the filter suitable for standard-definition video signals from all sources (e.g., broadcast and DVD). Broadcast video signals are channel limited: NTSC signals have 4.2 MHz bandwidth, and PAL signals have 5 MHz bandwidth. Video signals from a DVD player, however, are not channel limited; so the bandwidth of DVD video signals can approach the Nyquist limit of 6.75 MHz . Recommendation: ITU-R BT.601-5 specifies 13.5 MHz as the sampling rate for standarddefinition video. Therefore, the maximum bandwidth of the signal is 6.75 MHz . To ease the filtering requirements, most modern video systems oversample by two times, clocking the video current DAC at 27 MHz .

## Outputs

The video output amplifiers can both source and sink load current, allowing output loads to be DC- or ACcoupled. The amplifier output stage needs approximately 300 mV of headroom from either supply rail. The devices have an internal level-shift circuit that positions the sync tip at approximately 300 mV at the output.
If the supply voltage is greater than 3.135 V ( $5 \%$ below a 3.3 V supply), each amplifier can drive two DC-coupled video loads to ground. If the supply is less than 3.135 V , each amplifier can drive only one DC-coupled or AC-coupled video load.

## Applications Information

## Audio Switch for Dual SCART Connectors

In addition to video signals, SCART connectors also support left and right audio signals that are full duplex. Figure 1 shows a matching audio switch for the MAX9655. Notice that it can be made from low-cost, discrete components. It is assumed that the set-top box chip generates the left and right audio signals directly, or the set-top box chip sends an ${ }^{2}$ S stream to a stereo audio DAC that generates the left and right audio signals. In both cases, the audio signals are filtered and amplified by a dual audio op amp before they are presented to the audio switch.
Figure 2 shows a matching audio switch for the MAX9656. Similar to how the MAX9656 handles video signals, the audio signals from the set-top box chip or the audio signals from the TV SCART are routed to the VCR SCART.

## AC-Coupling the Outputs

The outputs can be AC-coupled since the output stage can source and sink current as shown in Figure 3. Coupling capacitors should be $220 \mu \mathrm{~F}$ or greater to keep the highpass filter, formed by the $150 \Omega$ equivalent resistance of the video transmission line, to a corner frequency of 4.8 Hz or below. The frame rate of PAL systems is 25 Hz . The corner frequency should be well below the frame rate.

## Low-Power Video Switches <br> for Dual SCART Connectors



Figure 1. Audio Switch for the MAX9655

## Low-Power Video Switches for Dual SCART Connectors



9S96XVW/GS96XVW

Figure 2. Audio Switch for the MAX9656

## Low-Power Video Switches for Dual SCART Connectors

Table 6. Quiescent and Average Power Consumption for MAX9655/MAX9656

| MEASUREMENTS | POWER <br> CONSUMPTION <br> (mW) | CONDITIONS |
| :--- | :---: | :--- |
| Quiescent Power <br> Consumption | 69 | No load. |
| Average Power <br> Consumption | 175 (MAX9655) | $150 \Omega$ to ground on <br> each output. 50\% <br> flat field signal on <br> each input. |
|  | 200 (MAX9656) |  |

Power Consumption
The quiescent power consumption and average power consumption of the MAX9655/MAX9656 are very low because of the 3.3 V operation and low-power circuit design. Quiescent power consumption is defined when the MAX9655/MAX9656 are operating without loads and without any video signals.
Average power consumption represents the normal power consumption when the devices drive real video signals into real video loads. It is measured when the MAX9655/MAX9656 drive a $150 \Omega$ load to ground with a $50 \%$ flat field, which serves as a proxy for a real video signal.
Table 6 shows the quiescent and average power consumption of the MAX9655/MAX9656.

## Power-Supply Bypassing and Ground

The MAX9655/MAX9656 operate from a single-supply voltage down to 2.7 V , allowing for low-power operation. Bypass VDD to GND with a $0.1 \mu \mathrm{~F}$ capacitor. Place all external components as close as possible to the device.

Chip Information
PROCESS: BiCMOS

Typical Application Circuits


MAX9655/MAX9656

## Low-Power Video Switches for Dual SCART Connectors

_ Typical Application Circuits (continued)


## Low-Power Video Switches for Dual SCART Connectors



MAX9655/MAX9656

## Low-Power Video Switches <br> for Dual SCART Connectors

For the latest package outline information and land patterns, go to www.maxim-ic.com/packages

| PACKAGE TYPE | PACKAGE CODE | DOCUMENT NO. |
| :---: | :---: | :---: |
| 16 QSOP | E16-4 | $\underline{\mathbf{2 1 - 0 0 5 5}}$ |
| 20 QSOP | E20-1 | $\underline{\mathbf{2 1 - 0 0 5 5}}$ |



## Low-Power Video Switches for Dual SCART Connectors

Revision History

| REVISION <br> NUMBER | REVISION <br> DATE | PESCRIPTION <br> PHGES |  |
| :---: | :---: | :--- | :---: |
| 0 | $3 / 08$ | Initial release | - |
| 1 | $10 / 08$ | Removed future product reference from MAX9656, updated Shutdown Supply <br> Current parameter, updated Table 6 | $1,2,12$ |
| 2 | $11 / 08$ | Updated Shutdown Supply Current maximum value in EC table | 2 |

